

Sampling and Reconstruction

The sampling and reconstruction process

- Real world: continuous
- Digital world: discrete

Basic signal processing

- Fourier transforms
- The convolution theorem
- The sampling theorem

Aliasing and antialiasing

- Uniform supersampling
- Stochastic sampling

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Imagers = Signal Sampling

All imagers convert a continuous image to a discrete sampled image by integrating over the active “area” of a sensor.

$$R = \int_T \int_\Omega \int_A L(x, \omega, t) P(x) S(t) \cos \theta dA d\omega dt$$

Examples:

- Retina: photoreceptors
- CCD array

We propose to do this integral using Monte Carlo Integration

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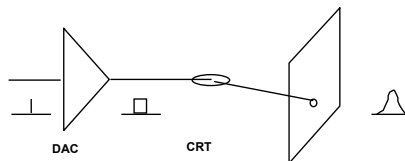
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Displays = Signal Reconstruction

All physical displays recreate a continuous image from a discrete sampled image by using a finite sized source of light for each pixel.

Examples:

- DACs: sample and hold
- Cathode ray tube: phosphor spot and grid

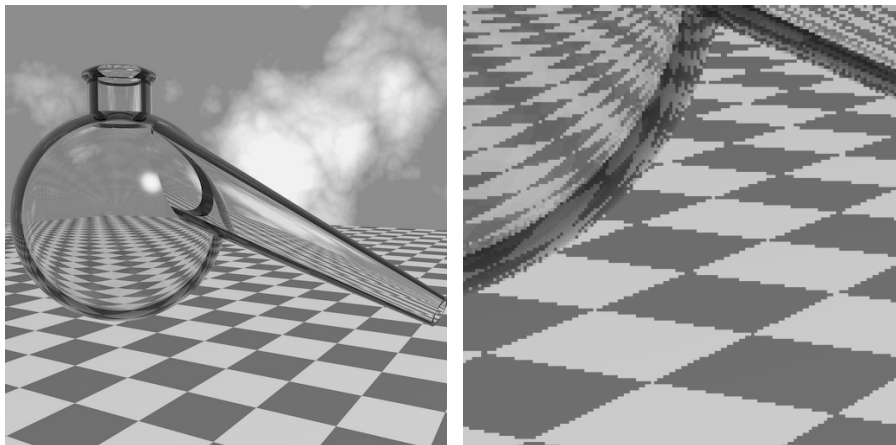


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Jaggies

Retort sequence by Don Mitchell



Staircase pattern or jaggies

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Sampling in Computer Graphics

Artifacts due to sampling - Aliasing

- **Jaggies**
- **Moire**
- **Flickering small objects**
- **Sparkling highlights**
- **Temporal strobing**

Preventing these artifacts - Antialiasing

Basic Signal Processing

Fourier Transforms

Spectral representation treats the function as a weighted sum of sines and cosines

Each function has two representations

- Spatial domain - normal representation
- Frequency domain - spectral representation

The *Fourier transform* converts between the spatial and frequency domain

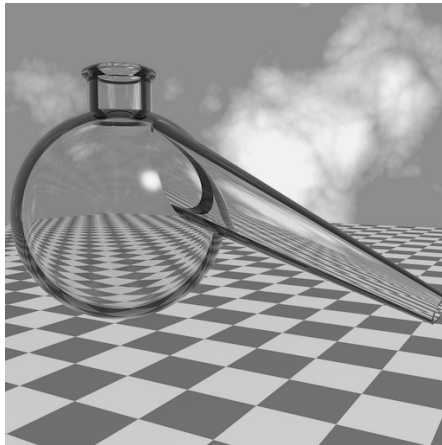
$$\begin{array}{ccc} \boxed{\text{Spatial Domain}} & \begin{array}{c} \Rightarrow F(\omega) = \int_{-\infty}^{\infty} f(x)e^{-i\omega x} dx \\ \Leftarrow f(x) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega)e^{i\omega x} d\omega \Leftarrow \end{array} & \boxed{\text{Frequency Domain}} \end{array}$$

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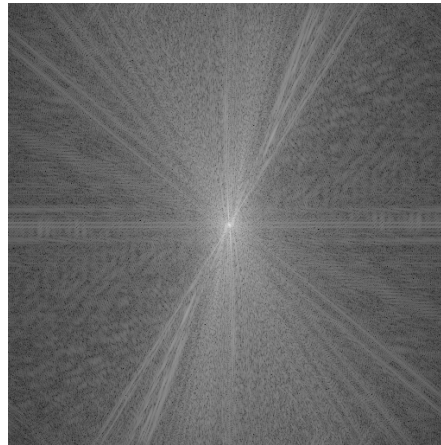
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Spatial and Frequency Domain

Spatial Domain



Frequency Domain

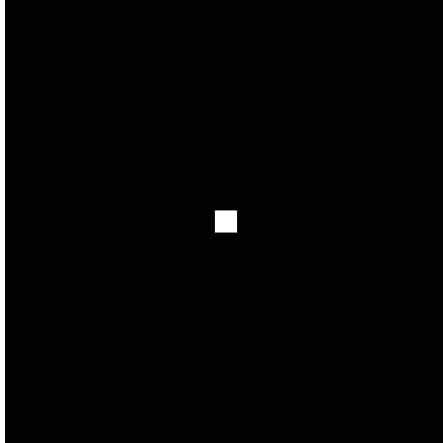


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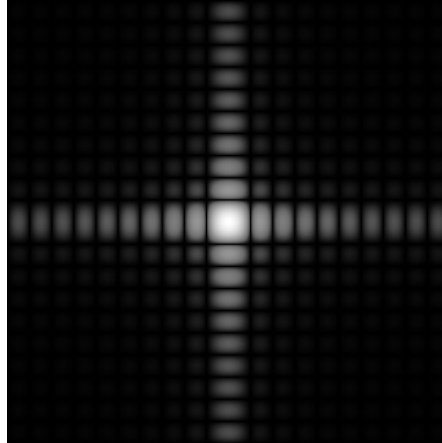
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More Examples

Spatial Domain



Frequency Domain

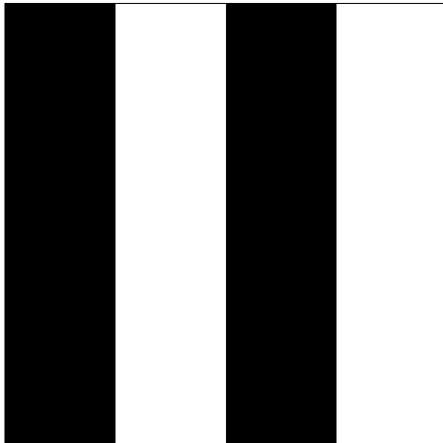


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More Examples

Spatial Domain



Frequency Domain

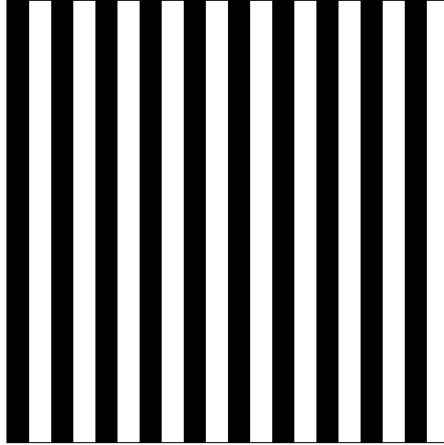


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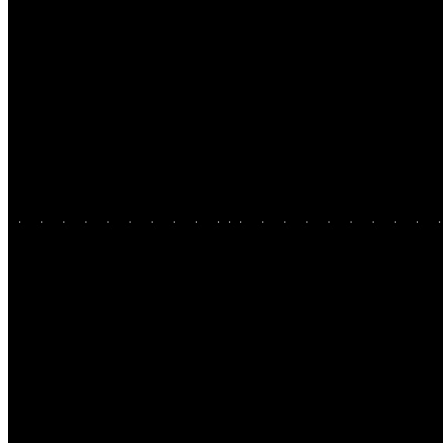
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More Examples

Spatial Domain



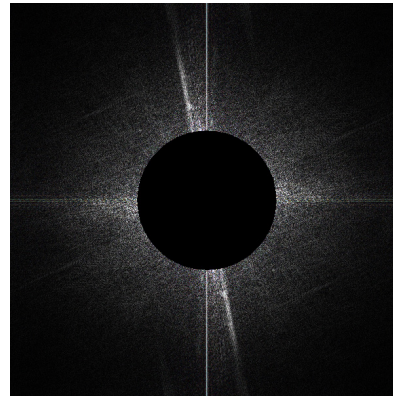
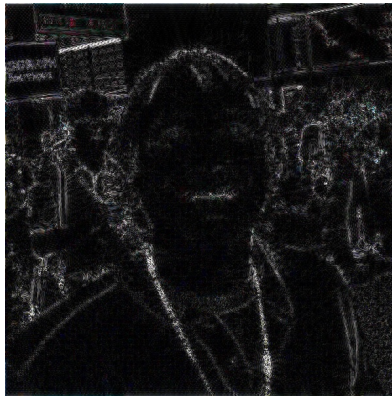
Frequency Domain



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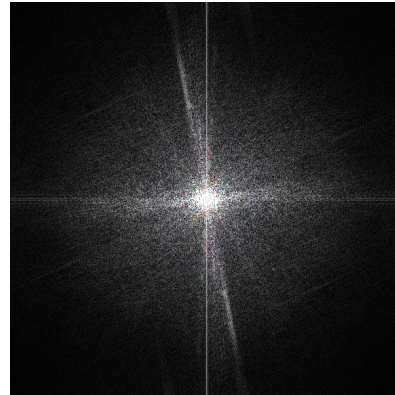
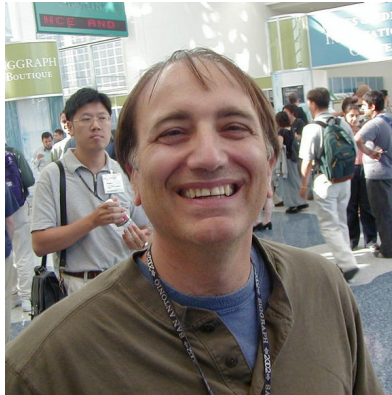
Pat's Frequencies



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Pat's Frequencies



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Convolution

Definition

$$h(x) = f \otimes g = \int f(x')g(x - x') dx'$$

Convolution Theorem: Multiplication in the frequency domain is equivalent to convolution in the space domain.

$$f \otimes g \leftrightarrow F \times G$$

Symmetric Theorem: Multiplication in the space domain is equivalent to convolution in the frequency domain.

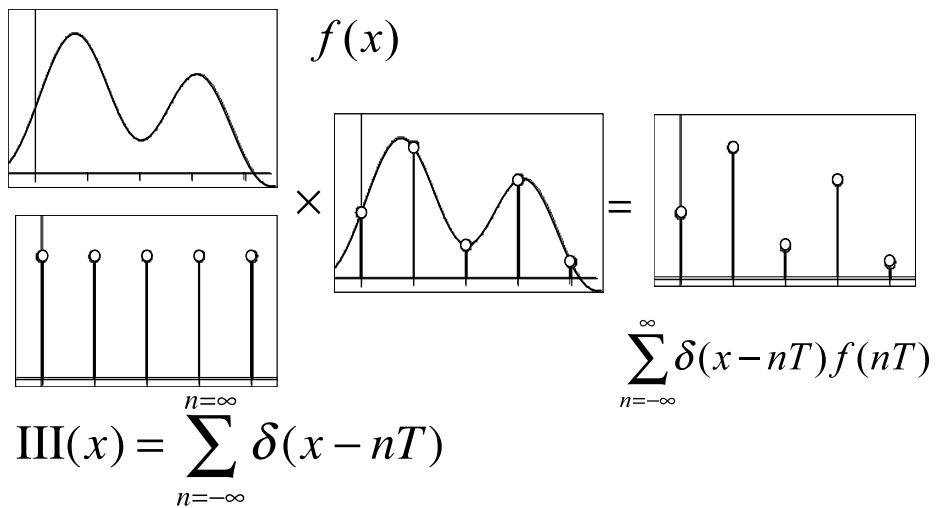
$$f \times g \leftrightarrow F \otimes G$$

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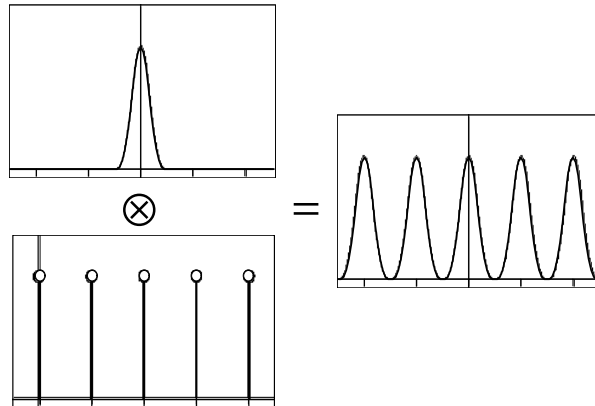
The Sampling Theorem

Sampling: Spatial Domain



Sampling: Frequency Domain

$F(\omega)$

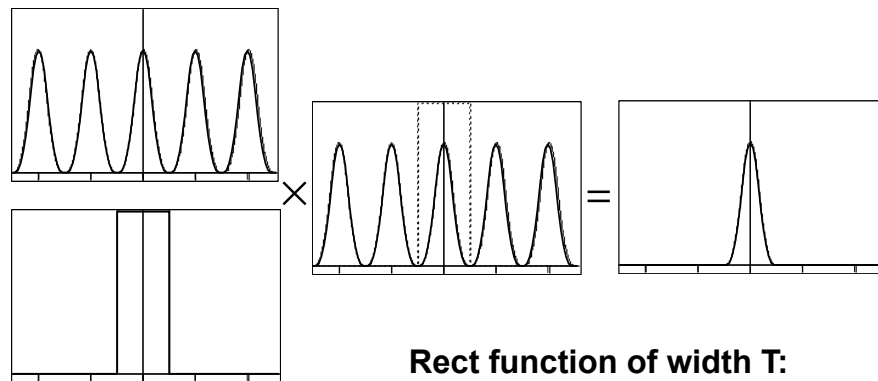


$$II_{1/T}(\omega) = \sum_{n=-\infty}^{\infty} \delta(\omega - n/T)$$

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Reconstruction: Frequency Domain



$$\Pi_{1/T}(x)$$

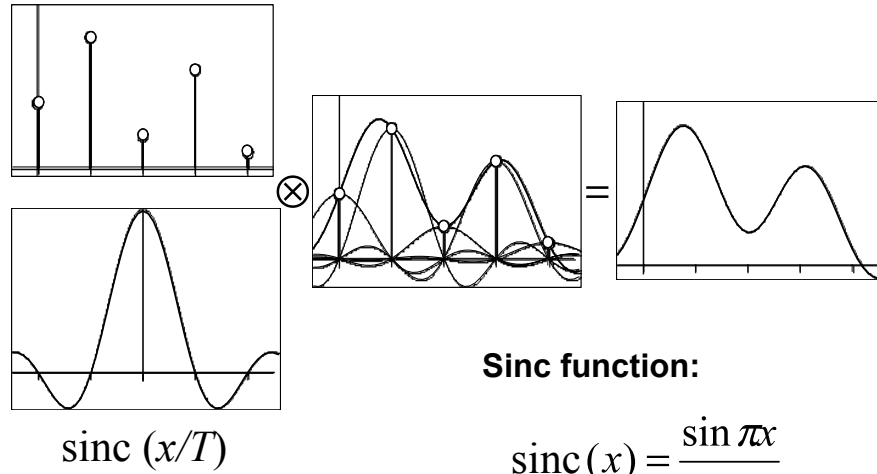
Rect function of width T:

$$\Pi_T(x) = \begin{cases} 1 & |x| \leq \frac{T}{2} \\ 0 & |x| > \frac{T}{2} \end{cases}$$

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Reconstruction: Spatial Domain



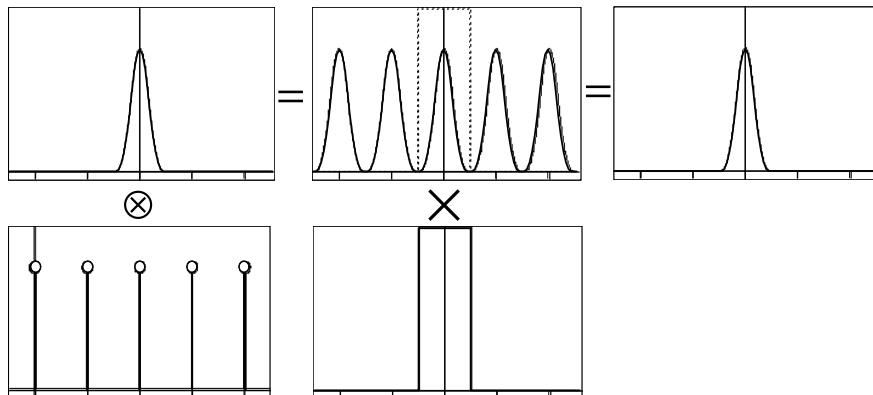
Sinc function:

$$\text{sinc}(x) = \frac{\sin \pi x}{\pi x}$$

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Sampling and Reconstruction



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Sampling Theorem

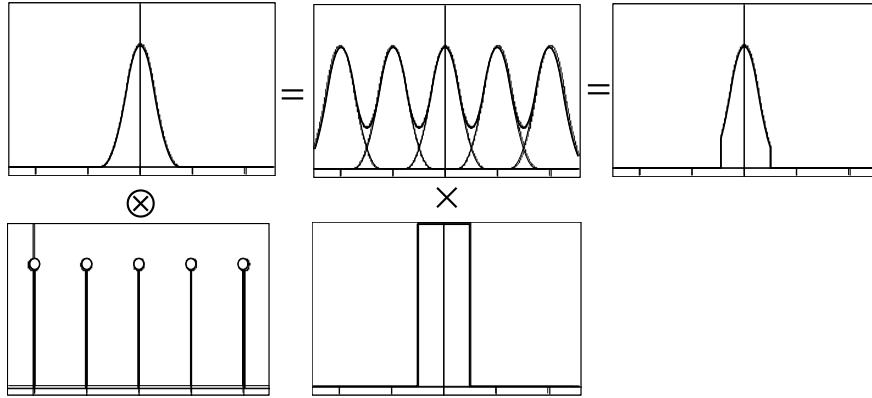
This result is known as the Sampling Theorem and is due to Claude Shannon who first discovered it in 1949

A signal can be reconstructed from its samples without loss of information, if the original signal has no frequencies above $1/2$ the Sampling frequency

For a given bandlimited function, the rate it must be sampled is called the *Nyquist Frequency*

Aliasing

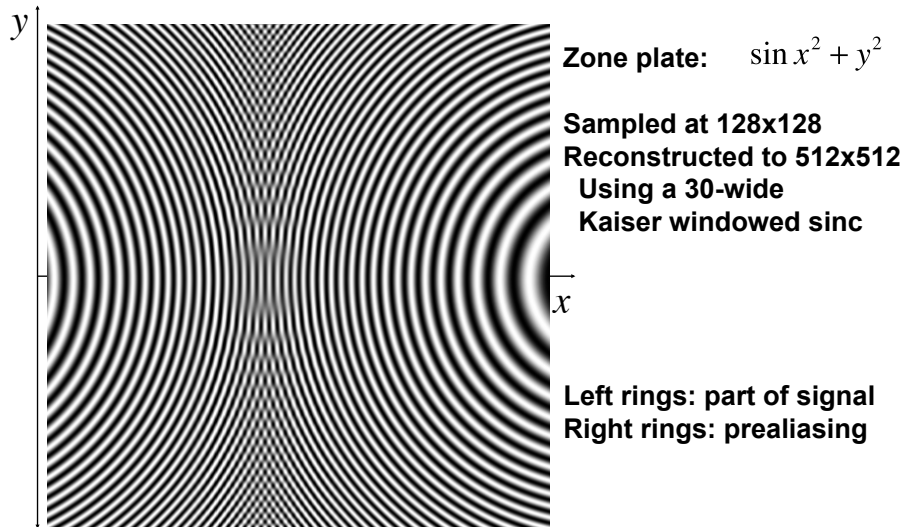
Undersampling: Aliasing



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Sampling a "Zone Plate"



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Ideal Reconstruction

Ideally, use a perfect low-pass filter - the sinc function - to bandlimit the sampled signal and thus remove all copies of the spectra introduced by sampling

Unfortunately,

- The sinc has infinite extent and we must use simpler filters with finite extents. Physical processes in particular do not reconstruct with sines
- The sinc may introduce ringing which are perceptually objectionable

Aliasing

- Prealiasing: due to sampling under Nyquist rate
- Postaliasing: due to use of imperfect reconstruction filter

Mitchell Cubic Filter

$$h(x) = \frac{1}{6} \begin{cases} (12 - 9B - 6C)x^3 + (-18 + 12B + 6C)x^2 + (6 - 2B) & |x| < 1 \\ (-B - 6C)x^3 + (6B + 30C)x^2 + (-12B - 48C)x + (8B + 24C) & 1 < |x| < 2 \\ 0 & \text{otherwise} \end{cases}$$

Good: (1/3, 1/3)

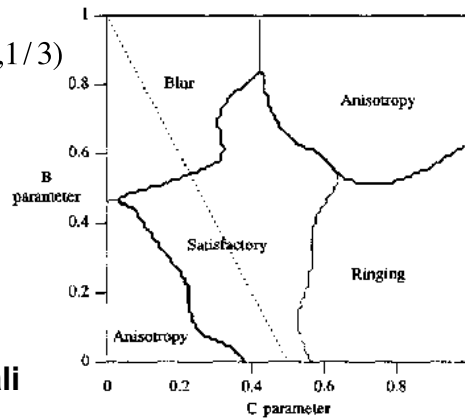
Properties:

$$\sum_{n=-\infty}^{n=\infty} h(x) = 1$$

B-spline: (1,0)

Catmull-Rom: (0, 1/2)

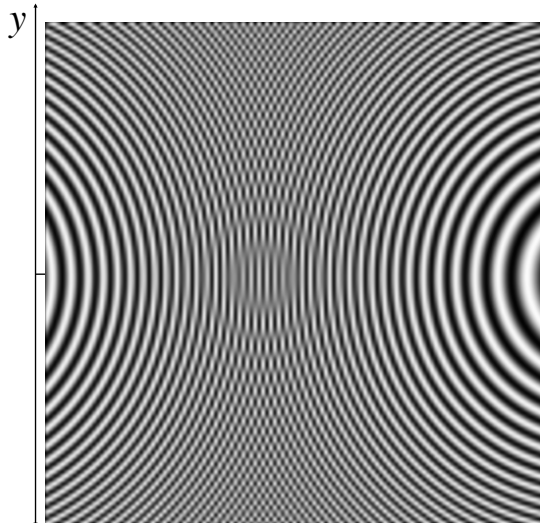
From Mitchell and Netravali



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Sampling a "Zone Plate"



Zone plate: $\sin x^2 + y^2$

Sampled at 128x128
Reconstructed to 512x512
Using optimal cubic

Left rings: part of signal
Right rings: prealiasing
Middle rings: postaliasing

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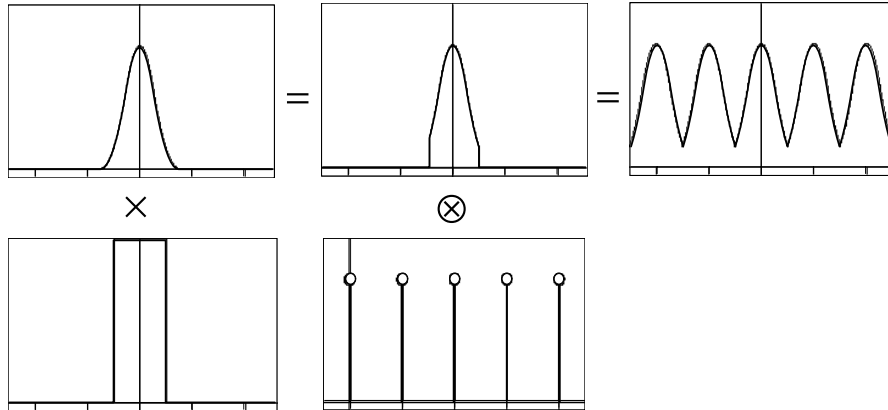
Antialiasing

Antialiasing

Antialiasing = Preventing aliasing

- 1. Analytically prefilter the signal**
 - Solvable for points, lines and polygons
 - Not solvable in general
 - e.g. procedurally defined images
- 2. Uniform supersampling and resample**
- 3. Nonuniform or stochastic sampling**

Antialiasing by Prefiltering



Frequency Space

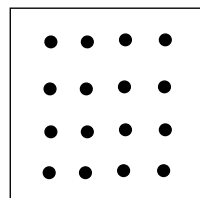
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Uniform Supersampling

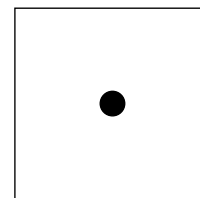
Increasing the sampling rate moves each copy of the spectra further apart, potentially reducing the overlap and thus aliasing

Resulting samples must be resampled (filtered) to image sampling rate



Samples

$$Pixel = \sum_s w_s \cdot Sample_s$$

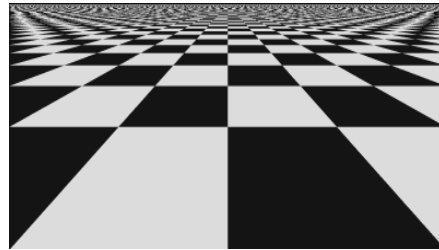
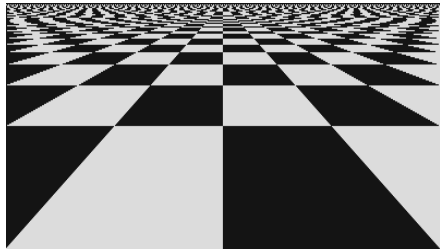
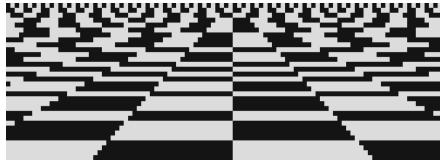


Pixel

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Point vs. Supersampled



Point

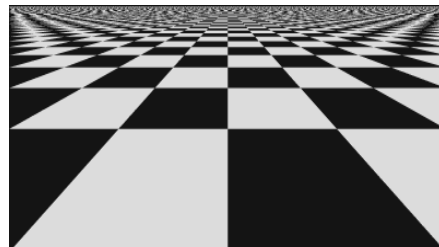
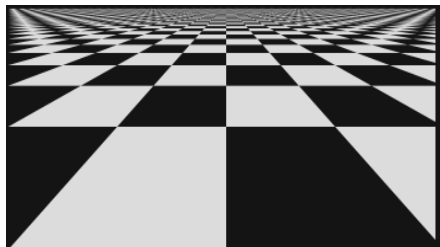
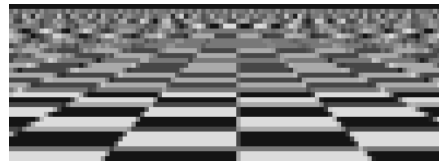
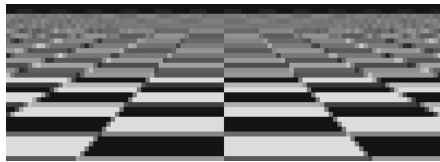
4x4 Supersampled

Checkerboard sequence by Tom Duff

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Analytic vs. Supersampled



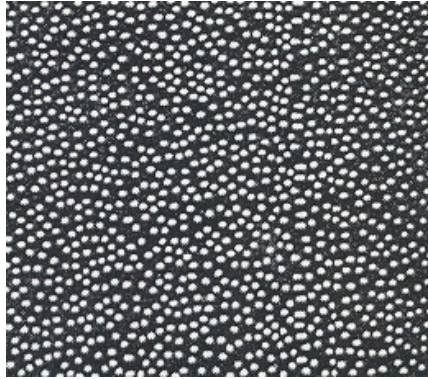
Exact Area

4x4 Supersampled

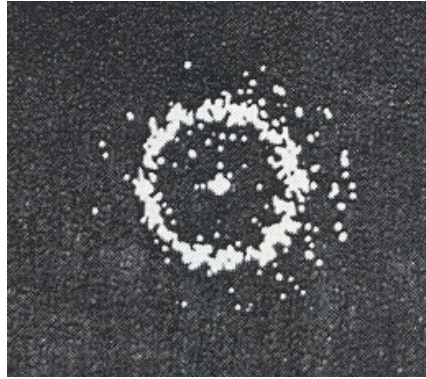
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Distribution of Extrafoveal Cones



Monkey eye
cone distribution



Fourier transform

Yellot theory

- Aliases replaced by noise
- Visual system less sensitive to high freq noise

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Non-uniform Sampling

Intuition

Uniform sampling

- The spectrum of uniformly spaced samples is also a set of uniformly spaced spikes
- Multiplying the signal by the sampling pattern corresponds to placing a copy of the spectrum at each spike (in freq. space)
- Aliases are coherent, and very noticeable

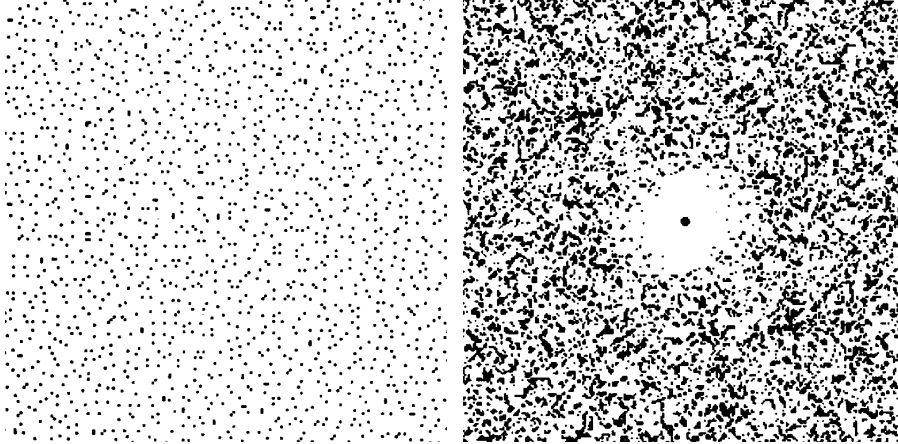
Non-uniform sampling

- Samples at non-uniform locations have a different spectrum; a single spike plus noise
- Sampling a signal in this way converts aliases into broadband noise
- Noise is incoherent, and much less objectionable

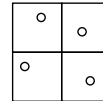
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Jittered Sampling



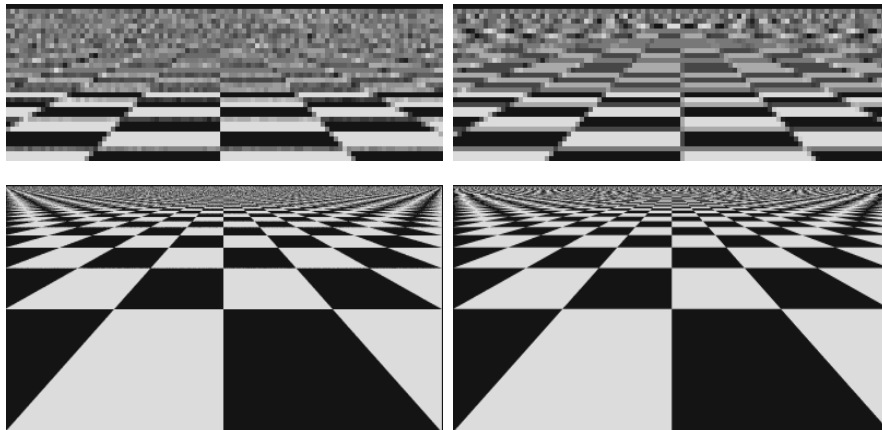
Add uniform random jitter to each sample



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Jittered vs. Uniform Supersampling



4x4 Jittered Sampling

4x4 Uniform

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Analysis of Jitter

Non-uniform sampling

$$s(x) = \sum_{n=-\infty}^{n=\infty} \delta(x - x_n)$$

$$x_n = nT + j_n$$

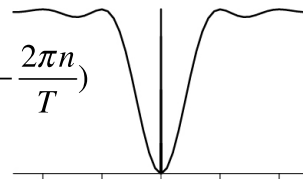
Jittered sampling

$$j_n \sim j(x)$$

$$j(x) = \begin{cases} 1 & |x| \leq 1/2 \\ 0 & |x| > 1/2 \end{cases}$$

$$J(\omega) = \text{sinc } \omega$$

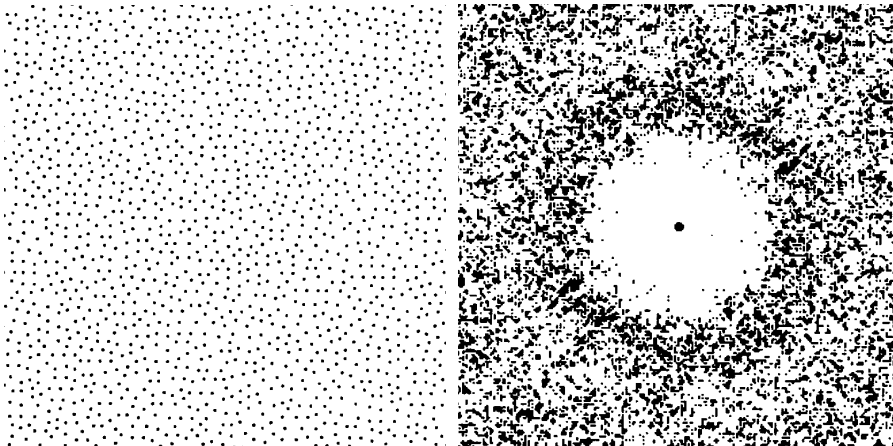
$$\begin{aligned} S(\omega) &= \frac{1}{T} \left[1 - |J(\omega)|^2 \right] + \frac{2\pi}{T^2} |J(\omega)|^2 \sum_{n=-\infty}^{n=\infty} \delta\left(\omega - \frac{2\pi n}{T}\right) \\ &= \frac{1}{T} \left[1 - \text{sinc}^2 \omega \right] + \delta(\omega) \end{aligned}$$



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Poisson Disk Sampling



Dart throwing algorithm

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